Decentralized Feedback Optimization via Sensitivity Decoupling: Stability and Sub-optimality

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Abstract-Online feedback optimization is a controller design paradigm for optimizing the steady-state behavior of a dynamical system. It employs an optimization algorithm as a dynamic feedback controller and utilizes real-time measurements to bypass knowing exact plant dynamics and disturbances. Different from existing centralized settings, we present a fully decentralized feedback optimization controller for networked systems to lift the communication burden and improve scalability. We approximate the overall input-output sensitivity matrix through its diagonal elements, which capture local model information. For the closed-loop behavior, we characterize the stability and bound the sub-optimality due to decentralization. We prove that the proposed decentralized controller yields solutions that correspond to the Nash equilibria of a noncooperative game. Simulations for a voltage control problem on a direct current power grid corroborate the theoretical results.

I. INTRODUCTION

Many engineering automation tasks consist of optimizing the steady-state operation of dynamical systems. Typical examples include congestion control in communication networks [1], voltage regulation [2], and optimal power flow in power systems [3]. The key is to select an appropriate control input to optimize an objective function that reflects the inputoutput performance. This pursuit, however, involves some critical obstacles. Real-world systems are often large-scale, networked, and complex, making it challenging to precisely model their behaviors. Furthermore, exogenous disturbances often appear as parameters in steady-state optimization problems. Since disturbances are difficult to measure in general, numerical optimization relying on an exact formulation can be prohibitive or require conservative approximations.

Emerging feedback optimization controllers [4] exhibit huge potential in regulating dynamical plants in an efficient manner. The central idea is to use optimization algorithms as feedback controllers to drive the plant to an optimal steadystate operating point. It utilizes real-time measurements to update control inputs without requiring the exact plant model and disturbances. This feature endows feedback optimization with the versatility to handle various scenarios. Examples include driving the system towards the global minimizer of convex functions [5], stationary points of non-convex functions [6], [7], and competitive equilibria (e.g., Nash or Wardrop) of noncooperative games [8], [9]. Guarantees on closed-loop stability, robustness, and constraint satisfaction are established in continuous-time [6], discrete-time [10], and sampled-data [8] scenarios.

The closed-loop interconnection between dynamical plants and feedback optimization controllers brings unique challenges. When the plants are rapidly pre-stabilized, they can be abstracted by their steady-state input-output maps [7], [10]. Such abstractions help to obtain closed-loop guarantees. However, for general dynamical plants [11], real-time output measurements may not align with steady-state responses. This mismatch may cause non-negligible errors in stability analysis. To address this issue, one can employ singular perturbation analysis to quantify the required timescale separation for a satisfactory feedback optimization controller [6]. The above methods require explicit knowledge of the steady-state input-output sensitivity matrix. To remove this restriction, recursive sensitivity learning based on streaming real-time measurements is proposed [12]. Moreover, zerothorder optimization algorithms utilizing gradient estimation [13] or Gaussian Processes [14] offer an alternative that circumvents the need for sensitivity information. It has proven effective in feedback optimization [11] and network optimization [15], [16].

Centralized approaches may encounter various issues (e.g., in scalability and privacy) when deployed in large-scale networked systems. In this regard, distributed implementations of feedback optimization have been studied [17]. These methods require that agents exchange local information with their neighbors in a network. To regulate the trade-off between global performance and local coordination, clustering strategies [18] explore different aspects (e.g., partitioning [19] and plug-and-play [20]) to efficiently control largescale networked systems. However, the establishment of communication channels may incur significant engineering effort, restrictions of feasible regions, time-consuming data transmission, and the need for iterative collaborative updates.

To overcome these limitations, we pursue a fully decentralized and communication-free approach. Each agent uses the local sensitivity information and updates its control input without any communication. From a network-level perspective, this corresponds to approximating the sensitivity matrix by its diagonal elements, thereby leading to fully decoupled updates. We first show that the stationary point of this controller coincides with the Nash equilibrium of an underlying convex game. Then, we conduct a comprehensive analysis of the closed-loop stability and sub-optimality, namely, the distance between the globally optimal point and the stationary point to which the proposed controller con-

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Fig. 1. Example of a system with three agents. The black line indicates the coupling dynamics between agents. The red dashed line represents the decentralized update using local inputs and outputs.

verges. We characterize the dependence of the sub-optimality on the dynamic coupling and the properties of the objective.

II. PRELIMINARIES AND PROBLEM FORMULATION

A. Notation

Let \mathbb{R} be the set of real numbers. We denote the inner product and the l_2 -norm by $\langle \cdot, \cdot \rangle$ and $\|\cdot\|$, respectively. For a differentiable function $\Phi : \mathbb{R}^n \times \mathbb{R}^n \to \mathbb{R}$, $\nabla_u \Phi(u, y)$ and $\nabla_y \Phi(u, y)$ denote the partial derivatives with respect to uand y, respectively. For a square matrix H, we use H_{diag} to represent the diagonal matrix with the diagonal elements of H. The largest singular value of H is $\sigma_{\max}(H)$, and the smallest one is $\sigma_{\min}(H)$. We denote the realization of a vector $x \in \mathbb{R}^n$ at time k by x_k , and its *i*-th entry by $x_{i,k}$.

B. Problem Formulation

In this paper, we focus on feedback optimization for a networked system with N agents. The plant is represented by an asymptotically stable linear time-invariant (LTI) system

$$x_{k+1} = Ax_k + Bu_k,$$

$$y_k = Cx_k + Du_k + d,$$
(1)

where $x \in \mathbb{R}^N$ is the state, $u \in \mathbb{R}^N$ is the input, $y \in \mathbb{R}^N$ is the output, and $d \in \mathbb{R}^N$ is a constant unknown disturbance. Every element of x, y, and u is the local state, output, and input of the corresponding agent. The plant (1) has a linear steady-state input-output map y = Hu + d, where $H = C(I - A)^{-1}B + D \in \mathbb{R}^{N \times N}$ is the sensitivity matrix. The scalar input and output of each agent are considered for notational convenience. In fact, the design and analysis can be easily extended to the multivariate setting with decoupled input constraint sets. An example of a system with three agents is given in Fig. 1.

We aim to drive the plant (1) to an optimal steady-state operating point defined by

$$\min_{u,y} \quad \Phi(u,y) \tag{2a}$$

s.t.
$$y = Hu + d$$
, (2b)

where the global objective function $\Phi(u, y) = \sum_{i=1}^{N} \Phi_i(u_i, y_i)$ is the sum of all the local objectives $\Phi_i(u_i, y_i)$, and u_i and y_i are the input and output of agent *i*, respectively. By replacing *y* in the objective, problem (2) is transformed to the following unconstrained problem:

$$\min_{u} \quad \tilde{\Phi}(u), \tag{3}$$

where $\tilde{\Phi}(u) = \sum_{i=1}^{N} \tilde{\Phi}_i(u)$, and $\tilde{\Phi}_i(u) = \Phi_i(u_i, y_i)$. We make the following assumptions about the objective function. Assumption 2.1: The following conditions hold:

- Agent i's objective function Φ_i(u_i, y_i) : ℝ×ℝ → ℝ is separable with respect to u_i and y_i, i.e., Φ_i(u_i, y_i) = Φ_i⁽¹⁾(u_i) + Φ_i⁽²⁾(y_i);
- 2) $\Phi_i^{(1)}(u_i) : \mathbb{R} \to \mathbb{R}$ is continuously differentiable, L_u -smooth, and m_u -strongly convex;
- 3) $\Phi_i^{(2)}(y_i) : \mathbb{R} \to \mathbb{R}$ is continuously differentiable, L_y -smooth, and m_y -strongly convex.

The assumptions on smoothness and strong convexity are common in the literature of numerical optimization and control [5], [21]. Furthermore, the assumption on separable objective functions is largely satisfied in practical applications [3]. It follows by Assumption 2.1 that the global objective functions $\Phi^{(1)}(u) = \sum_{i=1}^{N} \Phi_i^{(1)}(u_i)$ and $\Phi^{(2)}(y) =$ $\sum_{i=1}^{N} \Phi_i^{(2)}(y_i)$ are NL_u and NL_y -smooth, respectively. The objective function $\tilde{\Phi}(u)$ is *L*-smooth and *m*-strongly convex, where $L = NL_u + N\sigma_{\max}^2(H)L_y$ and $m = Nm_u + N\sigma_{\min}^2(H)m_y$, respectively.

III. DESIGN OF THE DECENTRALIZED CONTROLLER

Existing centralized feedback optimization controllers exploit the following gradient-based iterative update [4]

$$u_{k+1} = u_k - \eta \left(\nabla_u \Phi(u_k, y_k) + H^\top \nabla_y \Phi(u_k, y_k) \right), \quad (4)$$

where y_k is the real-time measurement, and $\eta > 0$ is the step size. Note that the update in (4) involves the full sensitivity matrix H, where its element H_{ij} represents the sensitivity of agent j's output to agent i's input. In practice, such a coupling sensitivity requires subsystems to exchange a large amount of information at each sampling instant k. In scenarios where i) communication is hard to establish (due to cost, privacy, etc.), ii) the sampling period is too short for communication to occur, or iii) the cross-coupling terms defining how agents influence one another are unknown, the centralized controller (4) may not be implementable in practice. Instead, we propose a fully decentralized controller via an approximate and decoupled sensitivity matrix.

A. Decentralized Controller

To develop a decentralized counterpart of (4), each agent updates its control input only based on its local sensitivity and measurement, i.e.,

$$u_{i,k+1} = u_{i,k} - \eta \left(\nabla_u \Phi_i(u_{i,k}, y_{i,k}) + H_{ii} \nabla_y \Phi_i(u_{i,k}, y_{i,k}) \right),$$

where the local sensitivity information H_{ii} and the local measurement $y_{i,k}$ are utilized to calculate $u_{i,k+1}$. From a system perspective, the update is represented by

$$u_{k+1} = u_k - \eta \left(\nabla_u \Phi(u_k, y_k) + H_{\text{diag}}^\top \nabla_y \Phi(u_k, y_k) \right).$$
(5)

By using the diagonal matrix H_{diag} to approximate the sensitivity matrix H, this controller is fully decoupled over the network. Hence, each agent can perform local adjustments without any communication.

B. A Game-Theoretic Interpretation

Next, we analyze the stationary points u_{∞} to which the decentralized controller (5) converges.

Theorem 3.1: Given Assumption 2.1, the stationary points of the controller (5), if they exist, equal to the Nash equilibria of the following convex game:

$$\forall i, \min \quad \Phi_i(u). \tag{6}$$

Proof: The *i*-th entry of the fixed point u_{∞} of (5) satisfies

$$\nabla_u \Phi_i(u_i, (Hu_\infty)_i + d_i) + H_{ii} \nabla_y \Phi_i(u_i, (Hu_\infty)_i + d_i) = 0.$$

The system-level representation of the above update is

$$\nabla_u \Phi(u_\infty, Hu_\infty + d) + H_{\text{diag}}^{\top} \nabla_y \Phi(u_\infty, Hu_\infty + d) = 0.$$

The left-hand side of the above equation is also the pseudogradient mapping, i.e., $u \mapsto [\partial_{u_1} \tilde{\Phi}_1(u), \cdots, \partial_{u_N} \tilde{\Phi}_N(u)]^\top$, of the game (6), whose zeros are the Nash equilibria [22, Theorem 5]. Thus, the stationary points of (5) are the Nash equilibria of the game (6).

IV. PERFORMANCE ANALYSIS

The update direction on the right-hand side of (5) does not align with the gradient of the objective function (3). To study the stability and sub-optimality of the interconnection of the controller (5) and the plant, we establish the condition of the plant such that the iteration of this decentralized controller is strongly monotone [23, Def. 22.1]. We develop the analysis for two separate cases, i.e., the plant is represented by the steady-state map (2b) or by the LTI plant (1).

A. Interconnection with an Algebraic Plant

We first consider the case where the plant has fastdecaying dynamics given any fixed inputs. Then, the plant can be replaced by its linear steady-state map (2b) [10]. The closed-loop system with the decentralized controller (5) is

$$y_k = Hu_k + d, \tag{7a}$$

$$u_{k+1} = u_k - \eta (\nabla \Phi^{(1)}(u_k) + H_{\text{diag}}^{\top} \nabla \Phi^{(2)}(y_k)).$$
 (7b)

Furthermore, the right-hand side of (7b) is strongly monotone when the plant (7a) is weakly coupled, see the characterization (8) below.

Lemma 4.1: Given Assumption 2.1, we conclude that $\nabla \Phi^{(1)}(u) + H_{\text{diag}}^{\top} \nabla \Phi^{(2)}(Hu+d)$ is a (m-c)-strongly monotone operator if m > c, where $m = Nm_u + N\sigma_{\min}^2(H)m_y$ and $c = N\sigma_{\max}(H - H_{\text{diag}})\sigma_{\max}(H)L_y$.

Proof: The proof can be found in Appendix A. Note that the condition m > c is equivalent to

$$\sigma_{\max}(H - H_{\text{diag}}) \le \frac{m_u + \sigma_{\min}^2(H)m_y}{\sigma_{\max}(H)L_y}.$$
 (8)

A diagonally dominant sensitivity matrix H can help to satisfy this condition.

If the plant satisfies (8), the stationary point u_{∞} of the closed-loop interconnection (7) exists and is unique [23, Corollary 23.37]. Let u^* be the globally optimal point of (2) and $y^* = Hu^* + d$ be the corresponding steady-state output.

The following theorem shows that the system (7) converges to a neighborhood of u^* with a linear rate.

Theorem 4.1: Let Assumption 2.1 hold, and let the plant (7a) satisfy the diagonal dominance (8). For all $\eta \in (0, \frac{2m-2c}{l^2-m^2})$, the closed-loop system (7) satisfies

$$\|u_{k}-u^{*}\| \leq \rho^{k} \|u_{0}-u^{*}\| + \eta \| (H^{\top}-H_{\text{diag}}) \nabla \Phi^{(2)}(y^{*})\| \sum_{j=0}^{k-1} \rho^{j},$$
(9)

where $\rho = \sqrt{1 - 2m\eta + L^2\eta^2} + c\eta$, $0 < \rho < 1$, and m and c are given in Lemma 4.1.

Proof: Please see Appendix B.

To characterize the sub-optimality, we bound the distance between the globally optimal point u^* and the limit point u_{∞} to which the controller (5) converges. The following theorem provides this upper bound.

Theorem 4.2: If 2m > 1 and the conditions in Theorem 4.1 hold, then we have

$$\|u^* - u_{\infty}\| \le \|(H^{\top} - H_{\text{diag}})\nabla\Phi^{(2)}(y_{\infty})\|\sqrt{\frac{1}{2m-1}}.$$
 (10)
Proof: Please see Appendix C.

We obtain the bound (10) by analyzing the trajectory of the centralized controller (4) while taking the limit point u_{∞} of the decentralized controller (5) as a reference. Note that 2m > 1 is a sufficient condition for establishing closed-loop stability from this perspective.

Remark 4.1: The sub-optimality of the controller (5) depends on the coupling degree $H^{\top} - H_{\text{diag}}$ and the properties of objective functions (e.g., $\nabla \Phi^{(2)}(y)$ and m). The bound decreases as m increases. It provides an estimate of the distance between the Nash equilibrium and the globally optimal solution. We will illustrate the tightness of this bound via simulations later in Section V.

B. Interconnection with an LTI Plant

Consider the closed-loop interconnection of the LTI plant (1) and the controller (5)

$$x_{k+1} = Ax_k + Bu_k,\tag{11a}$$

$$y_k = Cx_k + Du_k + d,\tag{11b}$$

$$u_{k+1} = u_k - \eta(\nabla \Phi^{(1)}(u_k) + H_{\text{diag}} \nabla \Phi^{(2)}(y_k)).$$
(11c)

The existence of plant dynamics will cause a difference between the real-time measurement and the steady-state response. Hence, we consider the coupled errors involving the distance to the steady state of the plant (1) given a fixed input (i.e., $||x - H_x u||^2$) and the distance to the equilibrium point of the controller (5) (i.e., $||u - u_{\infty}||^2$). These two coupled dynamics decay to zero with a linear rate if the steady-state map of the LTI plant (1) satisfies the diagonal dominance (8).

Theorem 4.3: Suppose that the stable LTI plant (1) satisfies (8) and that Assumption 2.1 holds. Then, there exists a positive constant η^* such that $\forall \eta \in (0, \eta^*)$, the closed-loop system (11) exhibits a linear convergence, i.e.,

$$\left\| \frac{x_k - H_x u_k}{u_k - u_\infty} \right\|^2 \le (\lambda_{\max}(\Xi))^k \left\| \frac{x_0 - H_x u_0}{u_0 - u_\infty} \right\|^2, \quad (12)$$

where $H_x = (I - A)^{-1}B$, $0 < \lambda_{\max}(\Xi) < 1$, Ξ is a symmetric matrix given in Appendix D, and η^* is given in Appendix D.

Proof: Please see Appendix D.

Note that the steady-state inputs of systems (11) and (7) are identical. By Lemma 4.1, if (8) holds, the stationary point u_{∞} of the system (7) is unique. It means that the system (11), if it converges, will converge to the stationary point of the system (7). Hence, the upper bound on the distance between u_{∞} and the globally optimal point u^* in (10) still holds.

V. NUMERICAL RESULTS

We consider a voltage control problem for a direct current (DC) power system. The model is adapted from [24], and it is illustrated in Fig. 2. This system consists of 8 nodes and 9 edges. There are two hub-like structures with a connection between nodes 4 and 5. The system dynamics are

$$\begin{bmatrix} C & 0 \\ 0 & L \end{bmatrix} \begin{bmatrix} \dot{V} \\ \dot{f} \end{bmatrix} = \begin{bmatrix} G & -B \\ B^{\top} & -R \end{bmatrix} \begin{bmatrix} V \\ f \end{bmatrix} + \begin{bmatrix} I^* + I_c \\ 0 \end{bmatrix}, \quad (13)$$
$$V_{\rm m} = V + d,$$

where $V \in \mathbb{R}^8$ is the node voltage, $V_m \in \mathbb{R}^8$ is the measured voltage with the unknown measurement error $d \in \mathbb{R}^8$, $f \in \mathbb{R}^8$ is the line current, $I^* \in \mathbb{R}^8$ is the reference current injection, and $I_c \in \mathbb{R}^8$ is the vector consisting of controllable current injection at each node. There are two diagonal matrices $C \in \mathbb{R}^{8\times8}$ and $G \in \mathbb{R}^{8\times8}$, where the diagonal elements represent the capacitance and the resistance of each node, respectively. Moreover, $L \in \mathbb{R}^{9\times9}$ and $R \in \mathbb{R}^{9\times9}$ are also diagonal matrices, where the diagonal elements indicate the inductance and the resistance of each line, respectively. The incidence matrix of the electrical network is $B \in \mathbb{R}^{8\times9}$.

Given an unknown change $\Delta I \in \mathbb{R}^8$ in current injection I^* , the objective is to operate the system such that the voltage measurement $V_{\rm m}$ at each node tracks the reference value $V_{\rm m, ref}$ with a minimal control effort related to I_c . Furthermore, we consider scenarios where communication is unfavorable (e.g., to avoid attacks). Hence, the current injection at each node is independently decided based on locally available measurements. This objective is

$$\min_{I_c, V_{\rm m}} \quad \frac{1}{2} (\gamma_1 \| I_c \|^2 + \gamma_2 \| V_{\rm m} - V_{\rm m, ref} \|^2)$$
s.t. $V_{\rm m} = H(I_c + I^* - \Delta I) + d,$
(14)

where the input-output steady-state sensitivity matrix $H = \begin{bmatrix} I & 0 \end{bmatrix} \begin{bmatrix} G & -B \\ B^{\top} & -R \end{bmatrix}^{-1} \begin{bmatrix} I \\ 0 \end{bmatrix} \in \mathbb{R}^{8 \times 8}$, and $V_{\rm m, \ ref} = HI^* + d$. We set C and L to be the identity matrix. We set I^* and ΔI to be all-ones vectors. The resistor R at each line equals 10. We choose different values of G to analyze cases with different levels of coupling (see (8)). The resulting system (13) is stable [24]. We further discretize the system (13) using Euler forward discretization with a step-size $\epsilon = 0.1$. We set $\gamma_1 = \gamma_2 = 1$ and select the step size used by the controllers $\eta = 0.05$, which meets the stability condition provided in Theorem 4.1.



Fig. 2. An 8-node DC power system.



Fig. 3. Relative distance to the globally optimal point (G=1).



Fig. 4. Relative distance to the globally optimal point with different values of G (i.e., different degrees of diagonal dominance).

In Fig. 3, the relative errors (i.e., $||u_k - u^*|| / ||u^*||$) for the input $u \triangleq I_c$ are plotted when G = 1. We observe that the normal gradient method (i.e., the centralized controller) brings the system to the optimal operating point, whereas the decentralized controller (5) leads to a sub-optimal solution. Furthermore, in Fig. 4, we illustrate the evolution of the upper bound (10) and the corresponding true sub-optimality when the parameter G varies from 1 to 100. As the degree of diagonal dominance of the sensitivity matrix H increases (i.e., G increases from 1 to 100), we observe a decrease in sub-optimality, which aligns with the results outlined in Theorem 4.2. Notably, the bound (10) provides a rather accurate estimation of the sub-optimality associated with the decentralized controller. From the perspective of the system operator, the bound (10) can act as a surrogate reference for the price of decentralization, namely, the extra cost incurred by disregarding the dynamic cross-coupling.

VI. CONCLUSION

We considered feedback optimization for networked systems, where inter-agent communication may be expensive or hard to establish. We proposed a fully decentralized controller through an approximated sensitivity matrix based on local model information. We proved that the resulting steady-state operating points are Nash equilibria. Moreover, we derived sufficient conditions for the stability of the closed-loop systems. We explicitly characterized the suboptimality of the decentralized controller, reflecting the price of decentralization. Through simulations on a DC power grid, we illustrated the performance of the decentralized controller and the tightness of the bound on sub-optimality.

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Appendix

A. Proof of Lemma 4.1

For
$$\forall u_1, u_2 \in \mathbb{R}^N$$
, we have

$$\begin{aligned} (\nabla \Phi^{(1)}(u_1) + H_{\text{diag}}^{\top} \nabla \Phi^{(2)}(Hu_1 + d) \\ &- \nabla \Phi^{(1)}(u_2) - H_{\text{diag}}^{\top} \nabla \Phi^{(2)}(Hu_2 + d))^{\top}(u_1 - u_2) \\ \stackrel{(\text{s.1})}{\geq} Nm_u \|u_1 - u_2\|^2 + \left(\nabla \Phi^{(2)}(Hu_1 + d) - \nabla \Phi^{(2)}(Hu_2 + d) \right)^{\top} \\ &\cdot (H_{\text{diag}}(u_1 - u_2) + Hu_1 + d - (Hu_2 + d) - H(u_1 - u_2)) \\ \stackrel{(\text{s.2})}{\geq} (Nm_u + Nm_y \sigma_{\min}^2(H)) \|u_1 - u_2\|^2 \\ &- \sigma_{\max}(H_{\text{diag}} - H) \sigma_{\max}(H) NL_y \|u_1 - u_2\|^2 \\ &= \left(Nm_u + Nm_y \sigma_{\min}^2(H) - \sigma_{\max}(H_{\text{diag}} - H) \sigma_{\max}(H) NL_y \right) \\ &\cdot \|u_1 - u_2\|^2 \\ &= (m - c) \|u_1 - u_2\|^2, \end{aligned}$$

where in (s.1), we use the strong convexity of $\Phi^{(1)}(u)$ and reorganize the terms. The strong convexity and the smoothness of $\Phi^{(2)}(y)$ are utilized in (s.2).

B. Proof of Theorem 4.1

The sub-optimality of the input u at time step k+1 satisfies

$$\begin{aligned} \|u_{k+1} - u^*\| &= \|u_k - u^* - \eta (\nabla \Phi(u_k) - (H^\top - H_{\text{diag}}) \\ &\cdot (\nabla \Phi^{(2)}(y_k) - \nabla \Phi^{(2)}(y^*) + \nabla \Phi^{(2)}(y^*))) \| \\ &\stackrel{(\text{s.1})}{\leq} \|u_k - u^* - \eta \nabla \tilde{\Phi}(u_k)\| + \eta \|(H^\top - H_{\text{diag}}) \nabla \Phi^{(2)}(y^*)\| \\ &+ \eta \sigma_{\max}(H^\top - H_{\text{diag}}) NL_y \sigma_{\max}(H) \|u_k - u^*\| \\ &\stackrel{(\text{s.2})}{\leq} (\sqrt{1 - 2m\eta + L^2 \eta^2} + c\eta) \|u_k - u^*\| \\ &+ \eta \|(H^\top - H_{\text{diag}}) \nabla \Phi^{(2)}(y^*)\|, \end{aligned}$$

where the triangle inequality and the smoothness of $\Phi^{(2)}(y)$ are utilized in (s.1). In (s.2), the strong convexity and smoothness of $\tilde{\Phi}(u)$ are used to bound $||u_k - u^* - \eta \nabla \tilde{\Phi}(u_k)||$. After reorganizing the terms, we obtain the above bound. Telescoping the inequality (s.2), we finished the proof with

$$\|u_k - u^*\| \le \rho^k \|u_0 - u^*\| + \eta \| (H^\top - H_{\text{diag}}) \nabla \Phi^{(2)}(y^*) \| \sum_{j=0}^{k-1} \rho^j,$$

where ρ is defined as in Theorem 4.1.

C. Proof of Theorem 4.2

Let $(\tilde{u}_k)_{k\in\mathbb{N}}$ and $\tilde{\eta}$ be the sequence of the inputs and the step size of the centralized controller (4), respectively. Then,

$$\begin{aligned} \|\tilde{u}_{k+1} - u_{\infty}\|^2 &= \|\tilde{u}_k - u_{\infty}\|^2 \\ &- 2\tilde{\eta}\nabla\tilde{\Phi}(\tilde{u}_k)^\top (\tilde{u}_k - u_{\infty}) + \tilde{\eta}^2 \|\nabla\tilde{\Phi}(\tilde{u}_k)\|^2 \end{aligned}$$

$$\begin{split} &\stackrel{(\mathrm{s},\mathrm{l})}{\leq} (1 - 2m\tilde{\eta} + L^2\tilde{\eta}^2) \|\tilde{u}_k - u_\infty\|^2 \\ &\quad + 2\tilde{\eta} \left((H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \right)^\top \\ &\quad \cdot (\tilde{\eta}(\nabla \tilde{\Phi}(\tilde{u}_k) - \nabla \tilde{\Phi}(u_\infty)) - (\tilde{u}_k - u_\infty)) \\ &\quad + \tilde{\eta}^2 \| (H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \|^2 \\ &\stackrel{(\mathrm{s},2)}{\leq} (1 - 2m\tilde{\eta} + L^2\tilde{\eta}^2) \|\tilde{u}_k - u_\infty\|^2 \\ &\quad + \tilde{\eta} \| (H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \|^2 \\ &\quad + \tilde{\eta} \| \tilde{\eta}(\nabla \tilde{\Phi}(\tilde{u}_k) - \nabla \tilde{\Phi}(u_\infty)) - (\tilde{u}_k - u_\infty) \|^2 \\ &\quad + \tilde{\eta}^2 \| (H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \|^2 \\ &\quad + \tilde{\eta}^2 \| (H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \|^2 \\ &\stackrel{(\mathrm{s},3)}{=} (1 - 2m\tilde{\eta} + L^2\tilde{\eta}^2) \| \tilde{u}_k - u_\infty \|^2 \\ &\quad + \tilde{\eta}(\| \tilde{\eta}(\nabla \tilde{\Phi}(\tilde{u}_k) - \nabla \tilde{\Phi}(u_\infty))^\top (\tilde{u}_k - u_\infty) + \| \tilde{u}_k - u_\infty \|^2) \\ &\quad + (\tilde{\eta}^2 + \tilde{\eta}) \| (H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \|^2 \\ &\stackrel{(\mathrm{s},4)}{\leq} (1 - 2m\tilde{\eta} + L^2\tilde{\eta}^2) (\tilde{\eta} + 1) \| \tilde{u}_k - u_\infty \|^2 \\ &\quad + (\tilde{\eta}^2 + \tilde{\eta}) \| (H^\top - H_{\mathrm{diag}}) \nabla \Phi^{(2)}(y_\infty) \|^2, \end{split}$$

where in (s.1), we rewrite $\nabla \tilde{\Phi}(\tilde{u}_k) = \nabla \tilde{\Phi}(\tilde{u}_k) - \nabla \tilde{\Phi}(u_{\infty}) + \nabla \tilde{\Phi}(u_{\infty}) - \nabla^* \tilde{\Phi}(u_{\infty})$ and reorganize the terms. In (s.2), we use the inequality $2a^{\top}b \leq ||a||^2 + ||b||^2, \forall a, b$. In (s.3), we expand $\|\tilde{\eta}(\nabla \tilde{\Phi}(\tilde{u}_k) - \nabla \tilde{\Phi}(u_{\infty})) - (\tilde{u}_k - u_{\infty})\|^2$ and reorganize the terms. In (s.4), we use the strong convexity and smoothness of $\tilde{\Phi}(u)$. By recursively using the above formula for $0, 1, \dots, k$, we obtain

$$\|\tilde{u}_{k} - u_{\infty}\|^{2} \leq \tilde{\rho}^{k} \|u_{0} - u_{\infty}\|^{2} + (\tilde{\eta} + \tilde{\eta}^{2}) \|(H^{\top} - H_{\text{diag}}) \nabla \Phi^{(2)}(y_{\infty})\|^{2} \sum_{j=0}^{k-1} \tilde{\rho}^{j},$$
(15)

where $\tilde{\rho} = (1 - 2m\tilde{\eta} + L^2\tilde{\eta}^2)(\tilde{\eta} + 1)$, and $\tilde{\rho} < 1$ for a sufficiently small $\tilde{\eta} > 0$ provided that 2m > 1. For every such $\tilde{\eta}$, (15) always holds, and the left-hand side of (15) remains the same. By analyzing the derivative of the right-hand side of (15) with respect to $\tilde{\eta}$, we observe that it is monotonically increasing for $\tilde{\eta} > 0$. When $\tilde{\eta} \to 0$, we obtain the smallest possible upper bound $||(H^{\top} - H_{\text{diag}})\nabla\Phi^{(2)}(y_{\infty})||^2 \frac{1}{2m-1}$.

D. Proof of Theorem 4.3

To start, we define

$$\varphi_1 = \nabla \Phi^{(1)}(u_k) + H_{\text{diag}} \nabla \Phi^{(2)}(y_k),$$

$$\varphi_2 = x_k - H_x u_k.$$

Then, we provide a preparatory lemma that presents some useful technical results. Note that $y_{\infty} \triangleq Hu_{\infty} + d$.

Lemma 4.1: The following inequalities hold:

1)
$$||x_{k+1} - H_x u_{k+1}||^2 \leq \eta^2 \varphi_1^\top H_x^\top H_x \varphi_1 + 2\eta \sigma_{\max}(H_x^\top A) ||\varphi_1|| ||\varphi_2|| + \varphi_2^\top A^\top A \varphi_2;$$

- 2) $\varphi_1^{\top}(u_k u_{\infty}) \geq Nm_u \|u_k u_{\infty}\|^2 + Nm_y \|y_k y_{\infty}\| NL_y \sigma_{\max}(H_{\text{diag}} H) \|y_k y_{\infty}\| \|u_k u_{\infty}\| NL_y \sigma_{\max}(C) \|y_k y_{\infty}\| \|\varphi_2\|;$
- 3) $\|y_k y_\infty\| \le \sigma_{\max}(C) \|\varphi_2\| + \sigma_{\max}(H) \|u_k u_\infty\|;$ 4) $\|\varphi_1\| \le (NL_u + NL_y \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(H)) \|u_k - u_\infty\| + NL_y \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(C) \|\varphi_2\|.$
- *Proof:* The proof can be found in our online report [25].

Now we are ready to prove Theorem 4.3. Consider

$$\begin{aligned} x_{k+1} - H_x u_{k+1} \\ u_{k+1} - u_{\infty} \end{aligned} \right\|^2 &= \|x_{k+1} - H_x u_{k+1}\|^2 + \|u_{k+1} - u_{\infty}\|^2 \\ &= \|x_{k+1} - H_x u_{k+1}\|^2 + \|u_k - u_{\infty}\|^2 \\ &- 2\eta \varphi_1^\top (u_k - u_{\infty}) + \eta^2 \|\varphi_1\|^2. \end{aligned}$$

By using the bounds derived in Lemma 4.1, we obtain

$$\left\| \frac{x_{k+1} - H_x u_{k+1}}{u_{k+1} - u_{\infty}} \right\|^2 \le \lambda_{\max}(\Xi) \left\| \frac{x_k - H_x u_k}{u_k - u_{\infty}} \right\|^2, \quad (16)$$

where $\Xi = \begin{bmatrix} \lambda_{\max}(A^{\top}A) + a_3\eta^2 + a_4\eta & a_1\eta^2 + a_2\eta \\ a_1\eta^2 + a_2\eta & 1 - m'\eta + L'\eta^2 \end{bmatrix}$ and the parameters therein are defined as

$$\begin{split} m' =& 2(Nm_u + Nm_y \sigma_{\min}^2(H) \\ &- NL_y \sigma_{\max}(H_{\text{diag}} - H) \sigma_{\max}(H)), \\ L' =& \lambda_{\max}(H_x^\top H_x + I)(NL_u + NL_y \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(H))^2, \\ a_1 =& \lambda_{\max}(H_x^\top H_x + I)NL_y \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(C) \\ &\cdot (NL_u + NL_y \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(H)), \\ a_2 =& \lambda_{\max}(H_x^\top A)(NL_u + NL_y \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(H)) \\ &+ 2Nm_y \sigma_{\max}(C) \sigma_{\max}(H) \\ &+ NL_y \sigma_{\max}(C)(\sigma_{\max}(H_{\text{diag}} - H) + \sigma_{\max}(H)), \\ a_3 =& \lambda_{\max}(H_x^\top H_x + I)N^2L_y^2 \sigma_{\max}^2(H_{\text{diag}}) \sigma_{\max}^2(C), \\ a_4 =& 2(\lambda_{\max}(H_x^\top A) \sigma_{\max}(H_{\text{diag}}) \sigma_{\max}(C) \\ &- (Nm_y \sigma_{\min}^2(C) - NL_y \sigma_{\max}^2(C))). \end{split}$$

Note that if m' > 0 and $\eta < \frac{m'}{L'}$, by Schur's complement, the sufficient and necessary condition for $\lambda_{\max}(\Xi) < 1$ is

$$(a_3m' + 2a_1a_2 - a_4L')\eta^2 + (a_4m' + a_2^2 + tL')\eta - tm' < 0,$$

where $t = 1 - \lambda_{\max}(A^{\top}A)$. The previous inequality yields the following stability bounds

$$\eta < \begin{cases} \eta_1^* & \text{if } a_3m' + 2a_1a_2 - a_4L' > 0, \\ \eta_2^* & \text{if } a_3m' + 2a_1a_2 - a_4L' \le 0, \end{cases}$$

where η_1^* and η_2^* are defined as

$$\begin{split} \eta_1^* = & \frac{\sqrt{(a_4m' + a_2^2 + tL')^2 + 4tm'(a_3m' + 2a_1a_2 - a_4L')} - (a_4m' + a_2^2 + tL')}{2(a_3m' + 2a_1a_2 - a_4L')} \\ \eta_2^* = & \frac{tm'}{a_4m' + a_2^2 + tL'}. \end{split}$$

The proof is finished by recursively applying (16).